

## ROOT-LOCUS ANALYSIS

### Lecture 12: Examples of Root Locus Plots

Example: Given that

$$KG(s) = \frac{K}{s(s+2)(s+4)}$$

Sketch the root locus of  $1 + KG(s) = 0$  and compute the value of  $K$  that will yield a “dominant” second order behavior with a damping ratio,  $\zeta = 0.7$ .

We have  $n = 3$  and  $m = 0$ .

Open loop zero: none

Open loop poles:  $s = 0, -2, -4$

Rule 1: The loci start from  $K = 0$  at the OL poles

Rule 2: The loci end at  $K \rightarrow \infty$  at 3 infinite zeros at  $|\infty|$

Rule 3: The number of loci is 3, as  $n = 3$

Rule 4: Root loci are symmetrical with respect to real axis

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Rule 5: There are  $(3 - 0) = 3$  asymptotes. The angles of asymptotes are given by:

$$\theta_j = \frac{(2j+1)\pi}{3-0} \quad ; j = 0,1,2$$

$$= \frac{\pi}{3}, \frac{3\pi}{3}, \frac{5\pi}{3}$$

i.e.  $\theta_j = 60^\circ, 180^\circ, 300^\circ (-60^\circ)$

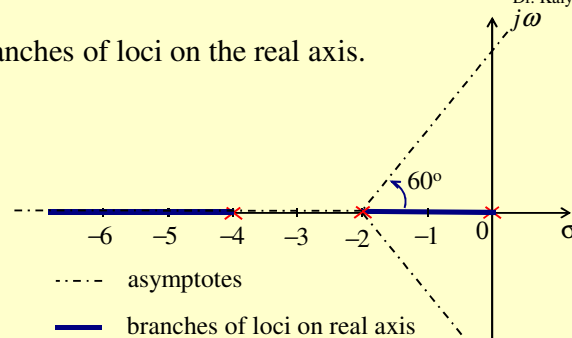
Rule 6: The point of intersection of the asymptotes is

$$\sigma_c = \frac{\sum OL\_poles - \sum OL\_zeros}{n - m}$$

$$= \frac{(0 - 2 - 4) - (-0)}{3 - 0} = -2.0$$

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Rule 7: Identify branches of loci on the real axis.



Rule 9: Crossing the imaginary axis.

The C.E. is  $s^3 + 6s^2 + 8s + K = 0$

$$\left( \frac{Y(s)}{R(s)} = \frac{K}{s^3 + 6s^2 + 8s + K} \right)$$

$s^3$	1	8	0
$s^2$	6	$K$	
$s^1$	$\frac{48-K}{6}$		
$s^0$	$K$		

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When  $K = 48$  in the  $s^1$  row we get an all zero row.

The Auxiliary Equation is

$$6s^2 + 48 = 0$$

$$s = \pm j2\sqrt{2}$$

Rule 10: The breakaway point is the solution of  $\frac{dK}{ds} = 0$

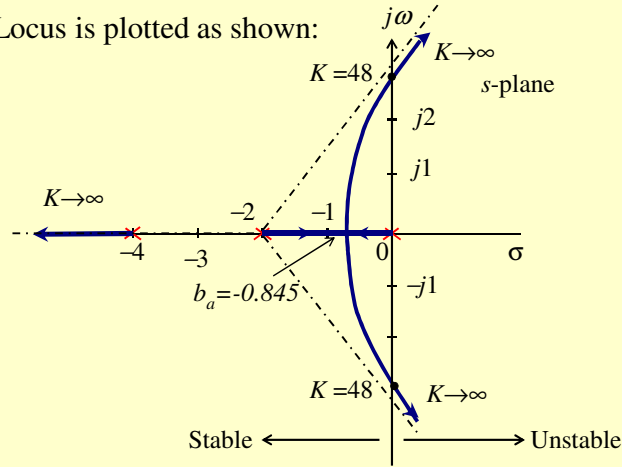
$$K = -(s^3 + 6s^2 + 8s)$$

$$\frac{dK}{ds} = -(3s^2 + 12s + 8) = 0$$

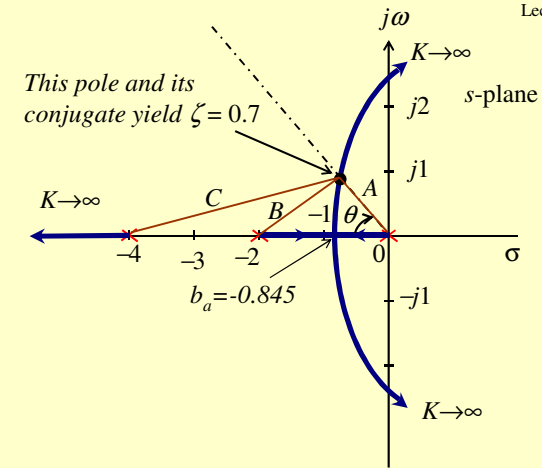
Solving,  $s = -0.845$  is the valid breakaway point for  $K > 0$ .

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The Root-Locus is plotted as shown:



With  $\zeta = 0.7$ , we have  $\theta = \cos^{-1} \zeta = 45.6^\circ$ . One of the poles that will give the required  $\zeta$  is given by the intersection of the root locus with the straight line that has angle  $\theta$  w.r.t. the negative real axis.



If the root-locus is sketched to scale, the value of  $K$  that will yield the required  $\zeta$  can be computed from the measured values of  $A$ ,  $B$  and  $C$ , i.e.

$$K_1 = A \times B \times C$$

**Example:** Construct the Root-Locus for a system with open-loop transfer function

$$KG(s)H(s) = \frac{K(s+3)}{s(s+5)(s+6)(s^2+2s+2)}$$

That is,  $n = 5$  and  $m = 1$ .

Open loop zero:  $s = -3$

Open loop poles:  $s = 0, -5, -6, -1 \pm j1$

**Rule 1:** The loci start from  $K = 0$  at the OL poles

**Rule 2:** The loci end at  $K \rightarrow \infty$  at the OL zero and 4 infinite zeros at  $|\infty|$

**Rule 3:** The number of loci is 5, as  $n = 5$

**Rule 4:** Root loci are symmetrical with respect to real axis

**Rule 5:** There are  $(5 - 1) = 4$  asymptotes. The angles of asymptotes are given by:

$$\theta_j = \frac{(2j+1)\pi}{4} \quad ; j = 0,1,2,3.$$

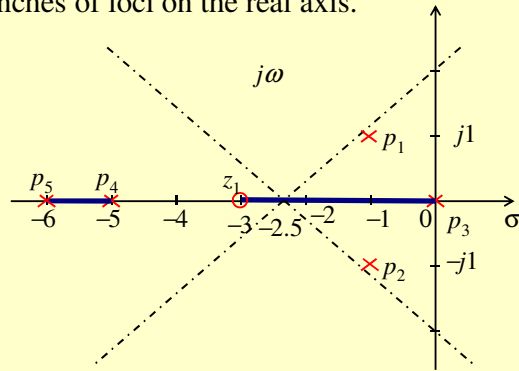
$$= \frac{\pi}{4}, \frac{3\pi}{4}, \frac{5\pi}{4}, \frac{7\pi}{4}$$

i.e.  $\theta_j = 45^\circ, 135^\circ, 225^\circ (-135^\circ), 315^\circ (-45^\circ)$

**Rule 6:** The point of intersection of the asymptotes is

$$\begin{aligned} \sigma_c &= \frac{\sum \text{OL\_poles} - \sum \text{OL\_zeros}}{n - m} \\ &= \frac{(0 - 5 - 6 - 1 + j1 - 1 - j1) - (-3)}{5 - 1} = -2.5 \end{aligned}$$

**Rule 7:** Identify branches of loci on the real axis.

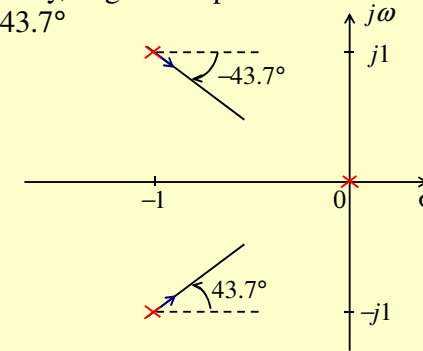


**Rule 8:** The angle of departure from the complex pole at  $(-1 + j1)$  is given by:

$$\begin{aligned} \theta_d &= 180^\circ + \left\{ \tan^{-1}\left(\frac{1}{3-1}\right) - \left[ \tan^{-1}(-1) + \tan^{-1}\left(\frac{1}{5-1}\right) + \tan^{-1}\left(\frac{1}{6-1}\right) + 90^\circ \right] \right\} \\ &= 180^\circ + \{ 26.6^\circ - [135^\circ + 14^\circ + 11.3^\circ + 90^\circ] \} \\ &= -43.7^\circ \end{aligned}$$

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By conjugate symmetry, angle of departure from the complex pole at  $(-1 - j1)$  is  $\theta_d = +43.7^\circ$



**Rule 9:** The points of intersection of root locus with the imaginary axis using the Routh-Hurwitz Criterion.

The Characteristic Equation is

$$s(s+5)(s+6)(s^2+2s+2) + K(s+3) = 0$$

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$$s^5 + 13s^4 + 54s^3 + 82s^2 + (60 + K)s + 3K = 0$$

$s^5$	1	54	$(60 + K)$	0
$s^4$	13	82	$3K$	0
$s^3$	47.7	$(60 + 0.77K)$	0	
$s^2$	$(65.6 - 0.21K)$	$3K$		
$s^1$	$\frac{3940 - 105K - 0.16K^2}{65.6 - 0.21K}$			
$s^0$	$3K$			

For stability,

$$\begin{aligned} 65.6 - 0.21K > 0 &\Rightarrow K < 309 \\ 3940 - 105K - 0.16K^2 > 0 &\Rightarrow K < 35 \\ 3K > 0 &\Rightarrow K > 0 \end{aligned}$$

i.e.  $0 < K < 35$

Substitute  $K = 35$  in the  $s^1$  row gives an all zero row.

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The Auxiliary Equation is

$$\begin{aligned} [65.6 - 0.21(35)]s^2 + 3(35) &= 0 \\ 58.2s^2 + 105 &= 0 \Rightarrow s = \pm j1.343 \end{aligned}$$

The root loci intersect the imaginary axis at  $s = \pm j1.343$

**Rule 10:** The breakaway point is the solution of  $\frac{dK}{ds} = 0$

$$\begin{aligned} K &= \frac{-(s^5 + 13s^4 + 54s^3 + 82s^2 + 60s)}{(s+3)} \\ \frac{dK}{ds} &= \frac{-(5s^4 + 52s^3 + 162s^2 + 164s + 60)}{(s+3)} + \frac{(s^5 + 13s^4 + \dots + 60s)}{(s+3)^2} = 0 \end{aligned}$$

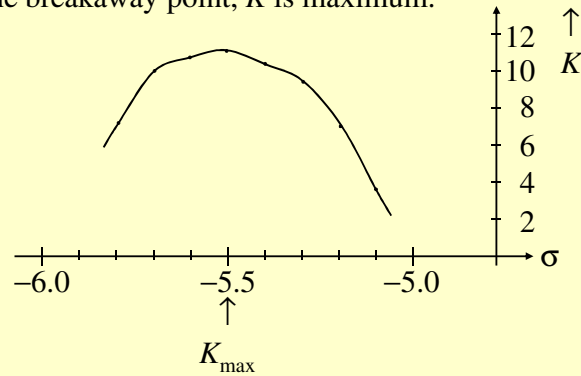
Solving,  $s = -5.52$  is the breakaway point.

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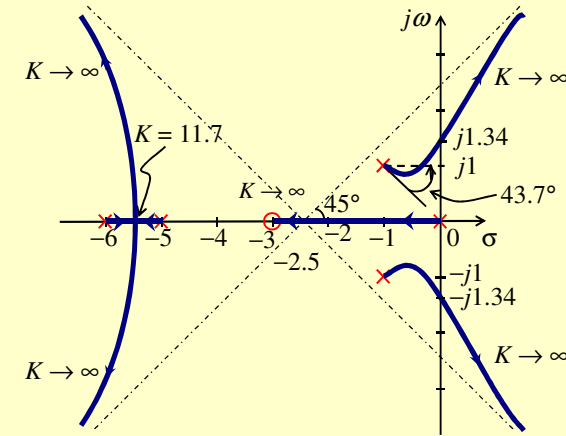
Alternatively, plot  $K$  vs  $s$

$s$	-5.1	-5.2	-5.3	-5.4	-5.5	-5.6	-5.7	-5.8
$K$	3.89	7.05	9.43	10.9	11.6	11.4	10.2	7.9

At  $s = -5.5$ , the breakaway point,  $K$  is maximum.



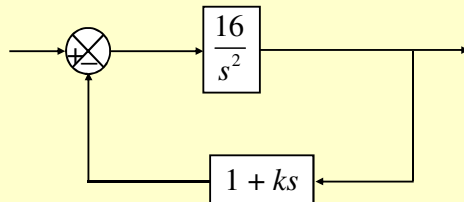
From the information obtained, the root-locus is sketched as shown:



### $K$ not a Multiplying Factor

If  $K$  is not a multiplying factor, some modification of the Characteristic Equation is required for constructing the root loci.

Example Consider the system



The Characteristic Equation for the system is

$$1 + \frac{16(1 + ks)}{s^2} = 0$$

Sketch the root locus w.r.t.  $k$ . From the root locus, compute the value of  $k$  that will give a damping ratio  $\zeta = 0.707$ .

Since  $k$  is not a multiplying factor, we modify the C.E. such that  $k$  appears as a multiplying factor.

The C.E. is  $s^2 + 16 + 16ks = 0$

Rewrite it as  $1 + \frac{16ks}{s^2 + 16} = 0$

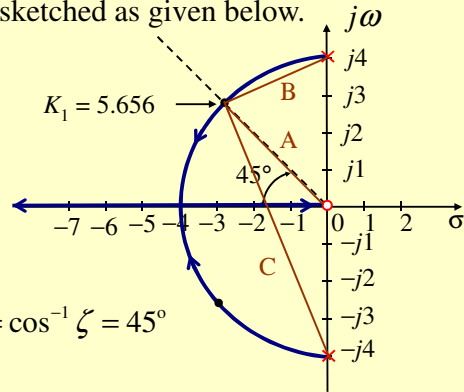
Define  $K = 16k$ , then the above equation becomes

$$1 + \frac{Ks}{s^2 + 16} = 0$$

Open loop poles:  $p_{1,2} = \pm j4.0$

Open loop zero:  $z_1 = 0$

The root locus can be easily sketched as given below.



With  $\zeta = 0.707$ , we have  $\theta = \cos^{-1} \zeta = 45^\circ$

From the root-locus, the value of  $K$  that will yield the required  $\zeta$  can be computed from the measured values of  $A$ ,  $B$  and  $C$ , i.e.

$$K_1 = \frac{B \times C}{A} = \frac{3.061 \times 7.391}{4} = 5.656$$

But  $K = 16k$ , hence  $k = 0.3535$ .

### Information Obtainable from the Root-Locus

- (1) The stability condition for any value of  $K$ , or other system parameter against which the root-locus is plotted.
- (2) The limits of  $K$  for which the system is stable.
- (3) The effect of variation of  $K$  on the performance potential of the system.
- (4) The actual characteristic equation for any value of  $K$ .
- (5) Evaluation of performance criteria such as undamped and damped natural frequencies, damping ratio and the transient response exponential terms.